57 Delay Optimisation in High-Performance Carry-Select Adders

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Abstract: This paper analyses techniques to measure the delay of 64 bit and 128-bit carry select adders. This is used for high-performance and low-power applications. It is introduced to work at a lower time delay than that required by a Ripple Carry Adder. This paper uses a very simple and efficient gate-level modification technique to significantly reduce the delay of the CSA. The proposed design has reduced delay and is compared with the 64 bit CSA. Xilinx ISE is used for simulation and synthesis.

Key words: RCA, CSA, delay, synthesis

INTRODUCTION

In current VLSI technology, adders have become important. The data processed by many digital systems may have delays. Design of area-efficient and high-speed data path logic systems are one of the most important areas of research in VLSI system design. In digital adders, the speed of addition is limited by the time required to propagate a carry through the adder. While Ripple Carry Adders (RCAs) have the most compact design (O(n) area) among all types of adders, they are the slowest types of adders (O(n) time). On the other hand, Carry Look-ahead Adders (CLAs) are the fastest adders (O(log(n)) time), but they are the worst from the area point of view (O(nlog(n)) area). Carry Select Adders (CSAs) have been considered as a compromise solution between RCAs and CLAs (O(n) time and O(2n) area) because they offer a good trade-off between the compact area of RCAs and the short delay of CLAs. As a result, some effort has been done to improve the efficiency of this CSA adder.

Design of the Carry Select Adder:

This work concentrates on static design styles since the performance advantage of both dynamic logic styles and pass-gate design is likely to decrease in deep-submicron technologies (European Semiconductor Industry Association, 2001 and Allam et al., 2000). The features of lesser dynamic power consumption and higher noise margin make static CMOS particularly attractive and useful (Yung et al., 2002 and De and Borkar, 2000). Moreover, the activation of the parasitic bipolar transistor is found to result in fatal erroneous states in dynamic logic and to make circuit design with pass-gates more tedious (Lu et al., 1997). The renewed interest in static design styles like pseudo-NMOS (Subba, 2000) and ratioed CMOS (Tretz, 2000) shows that alternative design styles are investigated in SOI in order to reduce the power dissipation while still maintaining high-speed performance. In the branch-based logic (BBL) design style, a logic cell is made of branches that contain some transistors in series (Masgonty, 1991). The branches are connected in parallel between the power supply VDD lines and the common output node. Many usual static CMOS gates have already a branch structure, as inverter and NAND / NOR gates. In the existing method, the 64-bit adder was split into four 16-bit sections (Amaury Nève et al., 2004). The true sum can be chosen by a multiplexer. The control signals for the multiplexers are the carry-in and the intermediate carry signals C15, C31 and C47. Fig 1 shows the intermediate carry signals and the carry-out are generated by C63 the carry-select boxes (CS-boxes). The inputs of the CS-boxes are the carry-in and the conditional carry signals and (with C0–15, C16–31, C32–47, or C48–63), which are all generated parallel (Hwang, 1979).

The notation C0-15 refers to the block carry signal for bit positions 0 to 15, assuming that the carry-in is at “0.” For the sake of preciseness, the indexes of the conditional carry signals have been simplified in Fig. 1. The carry-out and the intermediate carry signals are computed according to the equations presented in Table 1. To compute the carry-out, the intermediate carry is combined with and in one CS-C3 stage. At their turn, the 16-bit adder blocks are implemented as carry-select adders, with 4-bit adder blocks having a carry-in either at “0” or at “1.” At the 16-bit level, the same CS-boxes can be used as at the 64-bit level, sizing of the transistors being adapted to the particular load conditions. The 4-bit adder blocks can then finally be implemented as ripple carry adders, or also as carry-select adders, which was chosen here. The carry-select architecture is thus used at three different levels: in the 64-bit, the 16-bit, and the 4-bit adders (Lee et al., 2001, Kim et al., 2002, Bai et al., 2002, Sastiak et al., 2000, Garg and Katoch., 2001 and Ning, 2000).
Proposed Method:
This method analyses methods to measure the delay of 128-bit carry-select adders intended for high-performance and low-power applications. In this paper, a methodology to minimize the delay of 64 bit carry-select adders is presented. The methodology involves the design of a 128-bit CSA (Carry Select Adder) that minimizes the delay by using optimized circuit blocks. The delay is measured using logic equations and circuit blocks, as shown in the table below:

<table>
<thead>
<tr>
<th>Logic Equation</th>
<th>Circuit Block</th>
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<tbody>
<tr>
<td>$C_{15} = C_{15}^{in} + C_{15}^{0}$</td>
<td>CS-C0</td>
</tr>
<tr>
<td>$C_{31} = C_{31}^{in} C_{16}^{1} + C_{0}^{0} C_{16}^{1} + C_{16}^{1}$</td>
<td>CS-C1</td>
</tr>
<tr>
<td>$C_{47} = C_{47}^{in} C_{16}^{1} C_{32}^{1} + C_{0}^{0} C_{16}^{1} C_{32}^{0} + C_{16}^{0} C_{32}^{1} + C_{32}^{0}$</td>
<td>CS-C2</td>
</tr>
<tr>
<td>$C_{63} = C_{63}^{in} C_{16}^{1} C_{48}^{1} + C_{0}^{0} C_{16}^{1} C_{48}^{0} + C_{16}^{0} C_{48}^{1} + C_{48}^{0}$</td>
<td>CS-C0</td>
</tr>
</tbody>
</table>
select adders for high performance and low-power applications, by working at three levels of abstraction: design style, cell arrangement, and adder structure is dealt (Zaker and Zahnd, 1993, Katopis et al., 1999 and Shahidi, 2002).

The approach presented in this paper can be extended towards 128-bit carry-select adders. However, the number of carry-select levels and the adder architecture may be different in order to obtain an efficient realization (Aller and Kroell, 1999, Nève et al., 2002, Yano et al., 1990 and Martin, 2000). Fig 2 shows the block diagram of 128-bit CSA with 16 different sections of adders. The multiplexers shown in the Fig 2 can be used for the selection of the carry signals. Fig 3 and Fig 4 shows Simulation result and Synthesis Report of 64-Bit CSA. Fig 5 and Fig 6 shows Simulation result and Synthesis Report of 128-Bit CSA.

Fig. 3: Simulation Result of 64- Bit CSA.

Fig. 4: Synthesis Report of 64- Bit CSA.

Fig. 5: Simulation result of 128- Bit CSA.
**Fig. 6:** Synthesis Report of 128-Bit CSA.

**Results for CSA:**

The delay for 128-bit CSA using Xilinx software is measured as 77.214ns. It is compared with 16, 32, 64 bit CSA’s. Table I shows Comparison Results of various CSA’s.

<table>
<thead>
<tr>
<th>NO. OF BITS in CSA</th>
<th>DELAY MEASURED</th>
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<tbody>
<tr>
<td>16 BIT CSA</td>
<td>16.706ns</td>
</tr>
<tr>
<td>32 BIT CSA</td>
<td>23.058ns</td>
</tr>
<tr>
<td>64 BIT CSA</td>
<td>49.371ns</td>
</tr>
<tr>
<td>128 BIT CSA</td>
<td>77.214ns</td>
</tr>
</tbody>
</table>

**Table I:** Comparison Results of various CSA’s.

**Conclusion:**

A carry-select adder speeds 40% to 90% faster than RCA (whose delay is more) by performing additions in parallel and reducing the maximum carry path. Thus from the above research work it is concluded that carry select adder is one of the fastest adders that can be used in many addition operations. A simple approach can be proposed to reduce the area of SQRT CSLA architecture. The reduced number of gates of this work offers the great advantage in the reduction of area. The compared results show that the 128 bit CSA has a larger delay than 64 bit CSA. This work can be further extended to n-bit Carry Select Adder. The modified CSLA architecture is therefore, low area, simple and efficient for VLSI hardware implementations of LTI (Linear Time Invariant) system.

**REFERENCES**


